

A New Tuning Function-Based Robust Adaptive Controller for Parametric Strict-Feedback Systems

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The Adaptive Control Problem

Consider the system

$$\dot{x} = f(x, \theta, d) + g(x)u$$

affine in the uncertain parameter θ , where $d(t)$ is an unknown exogenous disturbance. Find a dynamic feedback

$$\begin{aligned} u &= u(x, \hat{\theta}) \\ \dot{\hat{\theta}} &= \tau(x, \hat{\theta}), \end{aligned}$$

where $\hat{\theta}$ is the estimate of θ , that drives the output to zero while keeping all closed-loop signals bounded.

Background

- In presence of bounded disturbances, performance of adaptive controllers can deteriorate and even lead to instability.
- Common solutions:
 - Robustifying (leakage) term: σ -modification (Ioannou, 1983) and e_1 -modification (Narendra, 1989).
 - Projection operators.
- **Leakage modifications** don't recover disturbance-free stability performance of unmodified adaptation law if disturbance disappears.
- **Projection operators** preserve ideal properties of adaptive controller if disturbance disappears, but require parameter bounds to be known a priori.

Background

- **Robust adaptive** controllers
 - Pomet and Praly (1992)
 - Zhang and Ioannou (1996)
 - Polycarpou and Ioannou (1996)
 - Freeman, Krstic, and Kokotovic (1998)
 - Ikhouane and Krstic (1998)
 - Marino and Tomei (1998)
 - Pan and Başar (1998)
 - Jiang and Hill (1999)
 - Ge and Wang (2003)
 - Cai, de Queiroz, and Dawson (2006)

Background

- Use of adaptive control was broadened with advent of the **integrator backstepping** design (KKK, 1995).
- Allows one to adaptively stabilize nonlinear systems in **parametric strict-feedback** (PSF) form.

$$\begin{aligned}\dot{x}_1 &= \varphi_1^\top(x_1) \theta + x_2 \\ \dot{x}_2 &= \varphi_2^\top(x_1, x_2) \theta + x_3 \\ &\vdots \\ \dot{x}_i &= \varphi_i^\top(x_1, \dots, x_i) \theta + x_{i+1} \\ &\vdots \\ \dot{x}_n &= \varphi_n^\top(x_1, \dots, x_n) \theta + u\end{aligned}$$

Background

- Standard backstepping
 - Recursive procedure generates at each step a new adaptation law \Rightarrow **Over-parametrization**.
 - n th-order PSF system ($n \geq 3$) requires the $(n - 2)$ th derivative of 1st adaptation law, the $(n - 3)$ th derivative of 2nd adaptation law, and so on.
 - Most projection operators are, at best, Lipschitz continuous (e.g., Pomet and Praly, 1992).
 - Sufficiently-smooth version of Pomet/Prally projection (Cai, de Queiroz, and Dawson, 2006).

Background

- When $d = 0$, **tuning functions** method (KKK, 1995) avoids differentiation of adaptation laws and overparametrization.
- When $d \neq 0$ and $\dot{\hat{\theta}} = \text{Proj}(x, \hat{\theta})$, tuning functions don't work.
 - Tuning functions in time derivative of Lyapunov-like function aren't cancelled by adaptation law nor can be proven nonpositive.
- Alternative: inject extra **nonlinear damping** terms to handle projection-related terms in time derivative of Lyapunov-like function (Marino and Tomei, 1998).
 - Avoids overparametrization.
 - High control effort.

Background

Common assumption: Disturbance is output of autonomous exosystem,

$$\dot{d} = s(d),$$

with partially known structure; e.g.,

$$s(d) = Sd$$

where S is an unknown stable matrix.

- Restricts class of disturbances.
- Facilitates control design.
 - Internal model principle to compensate for disturbance.
 - Observer to estimate disturbance.
 - No overparametrization.

Problem Statement

$$\begin{aligned}\dot{x}_1 &= \varphi_1^\top(x_1) \theta + x_2 + d_1 \\ &\vdots \\ \dot{x}_i &= \varphi_i^\top(x_1, \dots, x_i) \theta + x_{i+1} + d_i \\ &\vdots \\ \dot{x}_n &= \varphi_n^\top(x_1, \dots, x_n) \theta + d_n + u \\ y &= x_1\end{aligned}$$

$$x_i \in \mathbb{R}^{m_i}, i = 1, \dots, n$$

$\varphi_i \in \mathbb{R}^{p \times m_i}, i = 1, \dots, n$ are known nonlinearities

$\theta \in \mathbb{R}^p$ is an uncertain constant parameter vector

$d_i(t) \in \mathbb{R}^{m_i}, i = 1, \dots, n$ are unknown additive disturbances

Problem Statement

Assumptions:

- $\varphi_i \in \mathcal{C}^{n-i}$, $i = 1, \dots, n$.
- $d_i \in \mathcal{C}^0$, $i = 1, \dots, n$ and $\|d_i(t)\|_{\mathcal{L}_\infty} \leq \bar{d}_i$ where \bar{d}_i is unknown positive constant.
- Elements of θ satisfy $\underline{\theta}_i < \theta_i < \bar{\theta}_i$, $i = 1, \dots, p$ where $\underline{\theta}_i, \bar{\theta}_i$ are known bounds.
- Reference trajectory $y_r \in \mathcal{C}^n$ and $y_r^{(i)}(t) \in \mathcal{L}_\infty$, $i = 0, \dots, n + 1$.

Goal: Design a feedback $u(x_1, \dots, x_n, t)$ that ensures:

- $e(t) := y(t) - y_r(t) \rightarrow 0$ as $t \rightarrow \infty$, or $\|e(t)\| \rightarrow \varepsilon$ as $t \rightarrow \infty$ where ε can be made arbitrarily small.
- Boundedness of all closed-loop signals.

Preliminaries

Parameter transformation: Let

$$\bar{\theta} = (\bar{\theta}_1, \dots, \bar{\theta}_p)^\top, \quad \underline{\theta} = (\underline{\theta}_1, \dots, \underline{\theta}_p)^\top, \quad \Delta = \|\bar{\theta} - \underline{\theta}\|, \quad \tau = \frac{\bar{\theta} - \underline{\theta}}{\Delta}$$

then

$$\|\underline{\theta}\| < \|\theta\| < \|\bar{\theta}\| \quad \text{and} \quad \|\tau\| < 1$$

where $\|\cdot\|$ denotes the 2-norm.

Lemma: Consider the function

$$\begin{aligned} W = & \tau^\top \Gamma^{-1} \ln(\tau) + (\mathbf{1}_p - \tau)^\top \Gamma^{-1} \ln(\mathbf{1}_p - \tau) \\ & - \tau^\top \Gamma^{-1} \ln(\hat{\tau}) - (\mathbf{1}_p - \tau)^\top \Gamma^{-1} \ln(\mathbf{1}_p - \hat{\tau}) \end{aligned} \quad (1)$$

where $\hat{\tau} \in \mathbb{R}^p$ denotes the estimate of the transformed parameter τ , $\ln(\xi) := (\ln \xi_1, \dots, \ln \xi_p)^\top$, $\mathbf{1}_p := (1, \dots, 1)_{p \times 1}$, and Γ is a positive-definite diagonal matrix. Then, (1) is nonnegative and radially unbounded in

$$S := \{\hat{\tau} \in \mathbb{R}^p \mid 0 < \hat{\tau}_i < 1 \ \forall i\}.$$

Robust Adaptive Control Design

Design follows the tuning-function-based backstepping procedure.

Step 1

- Let $z_1 = e$, then

$$\dot{z}_1 = \varphi_1^T (\bar{\theta} - \tau \Delta) - \dot{y}_r + \alpha_1 + z_2$$

where $z_2 = x_2 - \alpha_1$.

- Design stabilizing function α_1 as

$$\alpha_1 = -c_1 z_1 - \varphi_1^T (\bar{\theta} - \hat{\tau} \Delta) + \dot{y}_r$$

where $c_1 > 0$.

Robust Adaptive Control Design

- Define function

$$V_1 = \frac{1}{2}z_1^2 + W(\hat{\tau}),$$

whose derivative along closed-loop z_1 -subsystem is

$$\dot{V}_1 = -c_1z_1^2 - \tilde{\tau}^\top \left(\varsigma_1 + \Gamma^{-1}N^{-1}\dot{\hat{\tau}} \right) + z_1z_2$$

where

$$\tilde{\tau} = \tau - \hat{\tau}, \quad \varsigma_1 = \varphi_1z_1\Delta, \quad N = \text{diag}[\hat{\tau}_i(1 - \hat{\tau}_i)]_{p \times p}.$$

Robust Adaptive Control Design

Step i ($2 \leq i \leq n - 1$)

- Let $z_i = x_i - \alpha_{i-1}$, then

$$\begin{aligned} \dot{z}_i &= \varphi_i^\top (\bar{\theta} - \tau \Delta) - \sum_{j=1}^{i-1} \frac{\partial \alpha_{i-1}}{\partial x_j} \left[\varphi_j^\top (\bar{\theta} - \tau \Delta) + x_{j+1} \right] \\ &\quad - \frac{\partial \alpha_{i-1}}{\partial \hat{\tau}} \dot{\hat{\tau}} - \sum_{j=1}^i \frac{\partial \alpha_{i-1}}{\partial y_r^{(j-1)}} y_r^{(j)} + \alpha_i + z_{i+1}. \end{aligned}$$

- Design stabilizing function α_i as

$$\begin{aligned} \alpha_i &= -c_i z_i - \varphi_i^\top (\bar{\theta} - \hat{\tau} \Delta) + \sum_{j=1}^{i-1} \frac{\partial \alpha_{i-1}}{\partial x_j} \left[\varphi_j^\top (\bar{\theta} - \hat{\tau} \Delta) + x_{j+1} \right] \\ &\quad + \sum_{j=1}^i \frac{\partial \alpha_{i-1}}{\partial y_r^{(j-1)}} y_r^{(j)} - z_{i-1} + \phi_i \end{aligned}$$

where $c_i > 0$ and ϕ_i is the **tuning function**.

Robust Adaptive Control Design

- Define function

$$V_i = V_{i-1} + \frac{1}{2}z_i^2,$$

whose derivative along closed-loop z_i -subsystem is

$$\begin{aligned} \dot{V}_i = & - \sum_{j=1}^i c_j z_j^2 - \tilde{\tau}^\top \left[\varsigma_i + \Gamma^{-1} N^{-1} \dot{\hat{\tau}} \right] \\ & - \sum_{j=1}^{i-2} \left(z_{j+1} \frac{\partial \alpha_j}{\partial \hat{\tau}} \right) N \Gamma \left[\varsigma_{i-1} + \Gamma^{-1} N^{-1} \dot{\hat{\tau}} \right] \\ & + z_i \left[\phi_i - \frac{\partial \alpha_{i-1}}{\partial \hat{\tau}} \dot{\hat{\tau}} \right] + z_i z_{i+1} \end{aligned}$$

where

$$\varsigma_i = \varsigma_{i-1} + \left(\varphi_i - \sum_{j=1}^{i-1} \frac{\partial \alpha_{i-1}}{\partial x_j} \varphi_j \right) z_i \Delta.$$

Robust Adaptive Control Design

- Set tuning function to

$$\phi_i = -\frac{\partial \alpha_{i-1}}{\partial \hat{\tau}} N \Gamma \varsigma_i + \bar{\phi}_i,$$

where $\bar{\phi}_i$ is an auxiliary tuning term.

$$\begin{aligned} \dot{V}_i = & -\sum_{j=1}^i c_j z_j^2 - \tilde{\tau}^\top \left[\varsigma_i + \Gamma^{-1} N^{-1} \dot{\hat{\tau}} \right] \\ & - \sum_{j=1}^{i-2} \left(z_{j+1} \frac{\partial \alpha_j}{\partial \hat{\tau}} \right) N \Gamma \left[\varsigma_{i-1} + \Gamma^{-1} N^{-1} \dot{\hat{\tau}} \right] \\ & - z_i \frac{\partial \alpha_{i-1}}{\partial \hat{\tau}} N \Gamma \left[\varsigma_i + \Gamma^{-1} N^{-1} \dot{\hat{\tau}} \right] + z_i \bar{\phi}_i + z_i z_{i+1}. \end{aligned}$$

Robust Adaptive Control Design

- Set auxiliary tuning function to

$$\bar{\phi}_i = - \sum_{j=1}^{i-2} \left(z_{j+1} \frac{\partial \alpha_j}{\partial \hat{\tau}} \right) N \Gamma \left(\varphi_i - \sum_{j=1}^{i-1} \frac{\partial \alpha_{i-1}}{\partial x_j} \varphi_j \right) \Delta$$

so that

$$\begin{aligned} \dot{V}_i = & - \sum_{j=1}^i c_j z_j^2 - \tilde{\tau}^\top \left[s_i + \Gamma^{-1} N^{-1} \dot{\hat{\tau}} \right] \\ & - \sum_{j=1}^{i-1} \left(z_{j+1} \frac{\partial \alpha_j}{\partial \hat{\tau}} \right) N \Gamma \left[s_i + \Gamma^{-1} N^{-1} \dot{\hat{\tau}} \right] + z_i z_{i+1}. \end{aligned}$$

Robust Adaptive Control Design

Step n

- Let $z_n = x_n - \alpha_{n-1}$, then

$$\begin{aligned}\dot{z}_n &= \varphi_n^\top (\bar{\theta} - \tau \Delta) - \sum_{j=1}^{n-1} \frac{\partial \alpha_{n-1}}{\partial x_j} \left[\varphi_j^\top (\bar{\theta} - \tau \Delta) + x_{j+1} \right] \\ &\quad - \frac{\partial \alpha_{n-1}}{\partial \hat{\tau}} \dot{\hat{\tau}} - \sum_{j=1}^n \frac{\partial \alpha_{n-1}}{\partial y_r^{(j-1)}} y_r^{(j)} + d_n + u.\end{aligned}$$

- Design control input

$$\begin{aligned}u &= -(c_n + k_n) z_n - \varphi_n^\top (\bar{\theta} - \hat{\tau} \Delta) \\ &\quad + \sum_{j=1}^{n-1} \frac{\partial \alpha_{n-1}}{\partial x_j} \left[\varphi_j^\top (\bar{\theta} - \hat{\tau} \Delta) + x_{j+1} \right] \\ &\quad + \sum_{j=1}^n \frac{\partial \alpha_{n-1}}{\partial y_r^{(j-1)}} y_r^{(j)} - z_{n-1} + \phi_n\end{aligned}$$

where $c_n, k_n > 0$ and ϕ_n is final tuning function.

Robust Adaptive Control Design

- Define function

$$V_n = V_{n-1} + \frac{1}{2}z_n^2,$$

whose derivative along closed-loop z_n -subsystem is

$$\begin{aligned} \dot{V}_n = & - \sum_{j=1}^n c_j z_j^2 - \tilde{\tau}^\top \left[s_n + \Gamma^{-1} N^{-1} \dot{\hat{\tau}} \right] \\ & - \sum_{j=1}^{n-2} \left(z_{j+1} \frac{\partial \alpha_j}{\partial \hat{\tau}} \right) N \Gamma \left[s_{n-1} + \Gamma^{-1} N^{-1} \dot{\hat{\tau}} \right] \\ & + z_n \left[\phi_n - \frac{\partial \alpha_{n-1}}{\partial \hat{\tau}} \dot{\hat{\tau}} \right] + z_n d_n - k_n z_n^2 \end{aligned}$$

where

$$s_n = s_{n-1} + \left(\varphi_n - \sum_{j=1}^{n-1} \frac{\partial \alpha_{n-1}}{\partial x_j} \varphi_j \right) z_n \Delta.$$

Robust Adaptive Control Design

- Set tuning function to

$$\phi_n = -\frac{\partial \alpha_{n-1}}{\partial \hat{\tau}} N \Gamma \varsigma_n - \sum_{j=1}^{n-2} \left(z_{j+1} \frac{\partial \alpha_j}{\partial \hat{\tau}} \right) N \Gamma \left(\varphi_n - \sum_{j=1}^{n-1} \frac{\partial \alpha_{n-1}}{\partial x_j} \varphi_j \right) \Delta$$

so that

$$\dot{V}_n = -\sum_{j=1}^n c_j z_j^2 - \tilde{\tau}^\top \left[\varsigma_n + \Gamma^{-1} N^{-1} \dot{\hat{\tau}} \right] - \sum_{j=1}^{n-1} \left(z_{j+1} \frac{\partial \alpha_j}{\partial \hat{\tau}} \right) N \Gamma \left[\varsigma_n + \Gamma^{-1} N^{-1} \dot{\hat{\tau}} \right] + z_n d_n - k_n z_n^2.$$

Robust Adaptive Control Design

- Design adaptation law as

$$\begin{aligned}\dot{\hat{\tau}} &= -N\Gamma\varsigma_n, \quad \hat{\tau}(0) \in \mathcal{S} \\ N &= \text{diag} [\hat{\tau}_i (1 - \hat{\tau}_i)]_{p \times p},\end{aligned}$$

which ensures $\hat{\tau}(t) \in \mathcal{S}$ for all time.

$$\dot{V}_n \leq - \sum_{j=1}^n c_j z_j^2 + \frac{d_n^2}{4k_n} \leq - \min_j (c_j) \|z\|^2 + \frac{\bar{d}_n^2}{4k_n}$$

where $z := (z_1, \dots, z_n)^\top$.

- V_n is nonnegative in $\mathbb{R}^n \times \mathcal{S}$ and \dot{V}_n is negative in the set

$$\left\{ z \in \mathbb{R}^n \mid \|z\| > \frac{\bar{d}_n}{2\sqrt{k_n \min_j (c_j)}} \right\}.$$

Robust Adaptive Control Design

- z is ultimately bounded, and bound can be made arbitrarily small.
- From Barbalat's lemma, if

$$d_n(t) \rightarrow 0 \quad \text{for } t \geq T > 0,$$

then

$$z(t) \rightarrow 0 \quad \text{as } t \rightarrow \infty.$$

Simulation Results

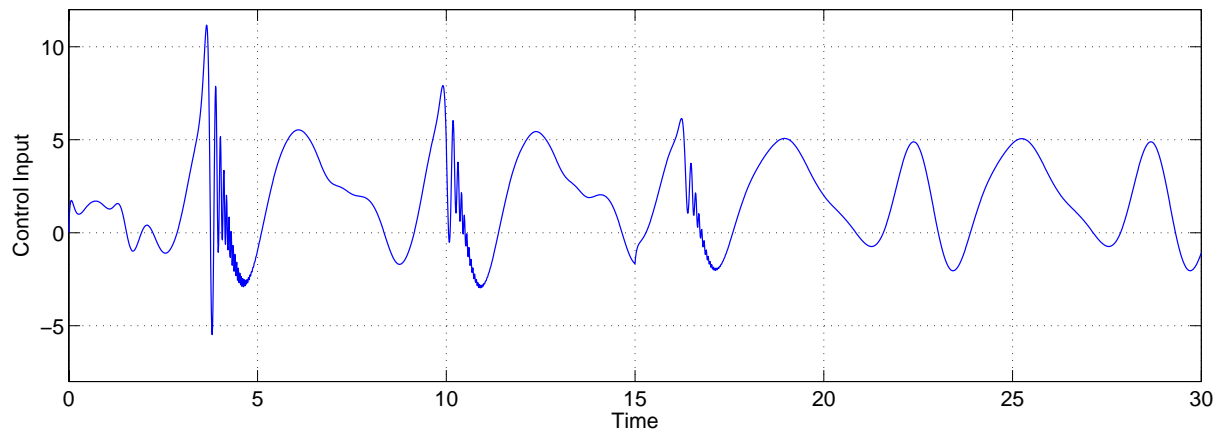
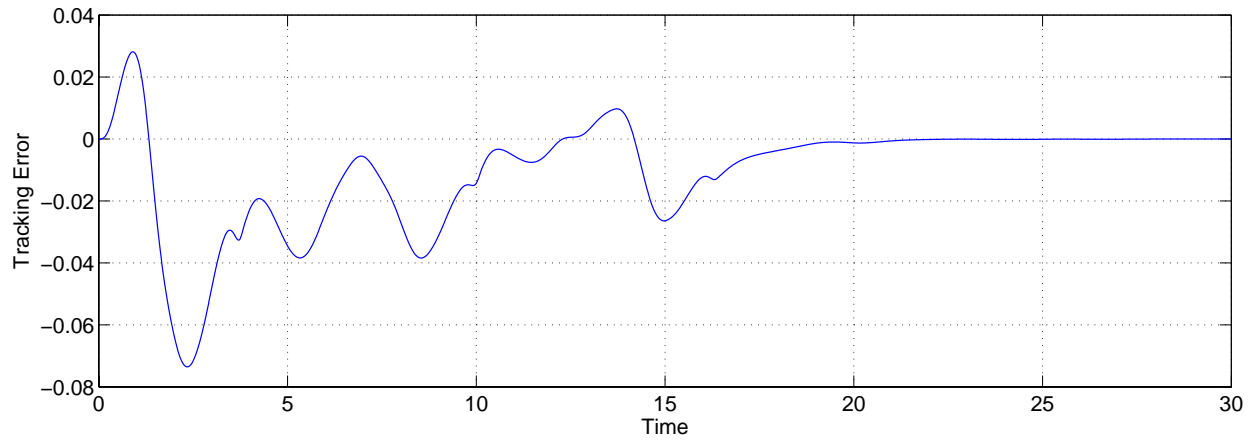
- System model

$$\begin{aligned}\varphi_1 &= \begin{bmatrix} -x_1^2 \\ 1 \\ \frac{1}{4}x_1^3 \end{bmatrix} & \varphi_2 &= \begin{bmatrix} -x_1^2 \\ 1 \\ \frac{1}{4}x_2^3 \end{bmatrix} & \varphi_3 &= \begin{bmatrix} -x_2 \\ 1 \\ \frac{1}{4}x_1^2x_3 \end{bmatrix} \\ \theta &= \begin{bmatrix} 2 \\ 1 \end{bmatrix} & \underline{\theta} &= \begin{bmatrix} 1.5 \\ 0.677 \end{bmatrix} & \bar{\theta} &= \begin{bmatrix} 3 \\ 2 \end{bmatrix} \\ d_3(t) &= \begin{cases} \sin 3t & 0 \leq t \leq 15 \\ 0 & \text{otherwise.} \end{cases}\end{aligned}$$

- Reference trajectory: $y_r = \sin t \left(1 - \exp \left(-\frac{1}{3}t^3 \right) \right)$
- Initial conditions: $x_1(0) = x_2(0) = x_3(0) = 0$, $\hat{\tau}(0) = (0.6, 0.4)^\top$
- Control parameters: $c_1 = 1$, $c_2 = 10$, $c_3 = 10$, $k_3 = 10$, $\Gamma = I_2$

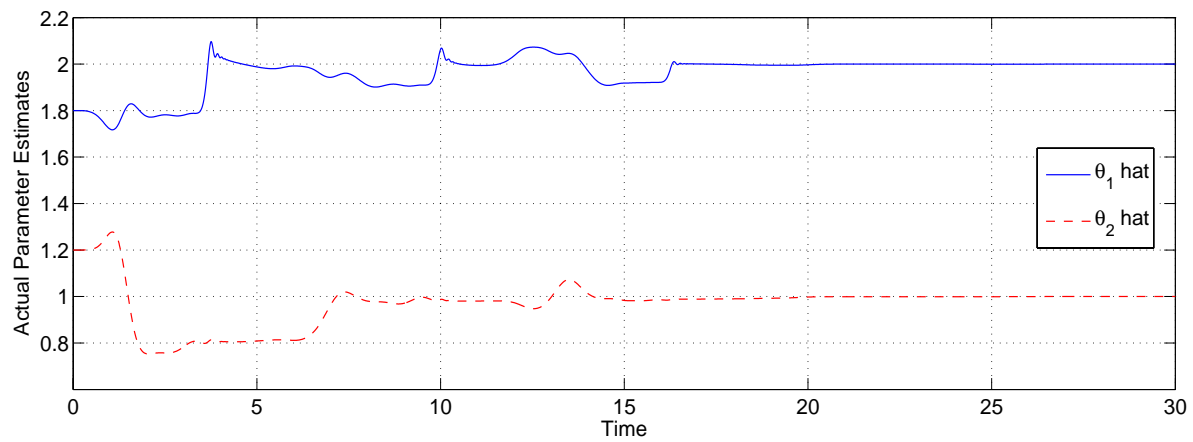
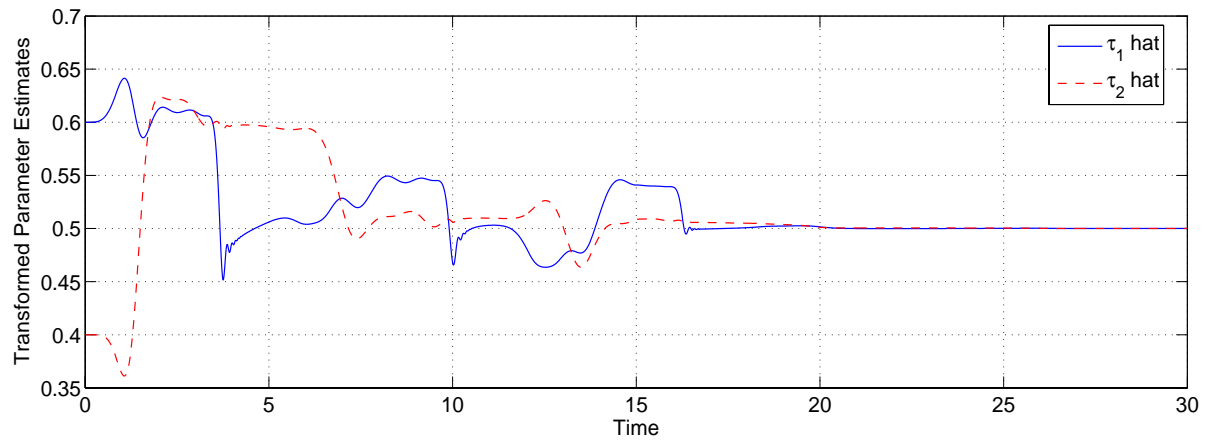
Simulation Results

- Tracking error $e(t)$ and control input $u(t)$



Simulation Results

- Parameter estimates $\hat{\tau}(t)$ and $\hat{\theta}(t)$



Concluding Remarks

- New robust adaptive controller for PSF systems.
- Additive disturbance is not exosystem generated.
- Only assumed disturbance is continuous and bounded.
- New projection-like operator.
 - Enabled use of tuning function method.
 - No overparametrization.
 - Avoided need for extra nonlinear damping terms.
- Practical tracking in presence of disturbance; asymptotic tracking when disturbance disappears.