# **Remarks on Tracking and Robustness Analysis for MEM Relays**



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Joint with Frédéric Mazenc and Marcio de Queiroz

2008 American Control Conference

# OUTLINE

- Types of Relays
- System Model and Reference Trajectories
- Error Dynamics and Objective
- Stability Theorem
- Crafting the Reference Trajectory
- Numerical Validation
- Conclusions

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- Voltage applied across electrodes creates an attractive capacitive or magnetic force between them.



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- In both types, pull-in occurs, i.e., the movable electrode suddenly 'crashes' onto the bottom electrode, so feedback control is necessary.



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- Zhu-Praly-.. '05: differential flatness, backstepping.
- Younis-Gao-de Queiroz '07: Lyapunov-based setpoint controller, feedback linearization tracking controller.

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upper (resp., lower) corresponds to electrostatic (resp., electromagnetic)

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x =position of the movable electrode, x = 0 in open position

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and 
$$\gamma = \begin{cases} 1/(\epsilon A) \\ R/(N\mu A) \end{cases}.$$

x = position of the movable electrode, x = 0 in open positionm = movable electrode mass, b = squeezed-film damping

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Electrostatic: q = charge

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Electrostatic:  $q = \text{charge}, \epsilon = \text{gap permittivity.}$ Electromagnetic: N = # coil turns.

# TRACKING

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Reference Trajectory:  $C^3 \ni y_d : [0, \infty) \to \mathbb{R}$  s.t.  $\exists$  constants  $m_i > 0$  for which

(a)  $m_1 \leq y_d(t) \leq m_2$ ,  $|\dot{y}_d(t)| \leq m_3$ , and  $|\ddot{y}_d(t)| \leq m_4 \ \forall t$ 

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# GOAL

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Reference trajectory, instead of a set point, can improve micro-relay performance.

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Setting  $e_1 = x - y_d$  and  $e_2 = \dot{e}_1$  and using the change of feedback  $u = \gamma (g_0 - x) z + \beta v_1 \sqrt{m/\alpha}$  gives

$$\begin{cases} \dot{e}_1 = e_2 \\ \dot{e}_2 = -\kappa_1 e_1 - \kappa_2 e_2 + \mu(e_1, e_2) + \zeta^2 + 2\zeta R_{\mu}(e_1, e_2, t) \\ \dot{\zeta} = v_1 - \frac{1}{2R_{\mu}(e_1, e_2, t)} \left\{ \ddot{\mathcal{Y}}_d(t) + \kappa_2 \ddot{\mathcal{Y}}_d(t) + \kappa_1 \dot{\mathcal{Y}}_d(t) + \dot{\mu} \right\} \end{cases}$$

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## **DESIRED STABILITY**

Goal: For each constant  $\mathcal{L} > 0$ , design  $\mu \in C^1$  and  $v_1$  for which:

G1 The closed loop  $Y = (e_1, e_2, \zeta)$  system is UGAS to 0.

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- G1 The closed loop  $Y = (e_1, e_2, \zeta)$  system is UGAS to 0.
- G2 There exist constant  $\underline{K}, \overline{K} > 0$  such that for all closed loop solutions with initial states  $Y(t_o) \in \underline{K}\mathcal{B}_3$ , we have

$$|Y(t)| \leq \overline{K}e^{-\mathcal{L}(t-t_o)}|Y(t_o)| \quad \forall t \geq t_o \geq 0.$$

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$$\mu(e_1, e_2) = -\frac{\kappa_1 m_1}{20} \left[ \sigma \left( \frac{20a_1}{\kappa_1 m_1} e_1 \right) + \sigma \left( \frac{20a_2}{\kappa_1 m_1} e_2 \right) \right].$$

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Then we can compute a constant  $\underline{a}(a_1, a_2)$  such that for each  $a_3 \geq \underline{a}$ ,

$$v_1 = -a_3\zeta(1+\zeta^2) + \frac{1}{2R_{\mu}(e_1, e_2, t)} \left\{ \ddot{y}_d(t) + \kappa_2 \ddot{y}_d(t) + \kappa_1 \dot{y}_d(t) \right\}$$

renders the error dynamics UGAS to the origin.

Theorem: Let  $a_1, a_2 > 0$  be given constants. Set  $\sigma(s) = s/\sqrt{1+s^2}$  and

$$\mu(e_1, e_2) = -\frac{\kappa_1 m_1}{20} \left[ \sigma \left( \frac{20a_1}{\kappa_1 m_1} e_1 \right) + \sigma \left( \frac{20a_2}{\kappa_1 m_1} e_2 \right) \right].$$

Then we can compute a constant  $\underline{a}(a_1, a_2)$  such that for each  $a_3 \geq \underline{a}$ ,

$$v_1 = -a_3\zeta(1+\zeta^2) + \frac{1}{2R_{\mu}(e_1, e_2, t)} \left\{ \ddot{y}_d(t) + \kappa_2 \ddot{y}_d(t) + \kappa_1 \dot{y}_d(t) \right\}$$

renders the error dynamics UGAS to the origin. Moreover, for each constant  $\mathcal{L} > 0$ , we can choose values of the  $a_i$ s and  $\underline{K}, \overline{K} > 0$  to satisfy G2.

## **IDEA of PROOF**

Construct explicit constants K and  $\Gamma$  so that the  $Y = (e_1, e_2, \zeta)$  dynamics has the strict global Lyapunov function

$$V_{3}(e_{1}, e_{2}, \zeta) = V_{2}(e_{1}, e_{2}) + \Gamma Q(\zeta), \text{ where}$$

$$V_{2}(e_{1}, e_{2}) = e_{1}e_{2} + KV_{1}(e_{1}, e_{2}),$$

$$Q(\zeta) = \frac{1}{a_{3}} \left(\frac{1}{2}\zeta^{2} + \frac{1}{4}\zeta^{4}\right),$$
and 
$$V_{1}(e_{1}, e_{2}) = \frac{1}{2}e_{2}^{2} + \int_{0}^{e_{1}} \left\{\kappa_{1}l + \frac{\kappa_{1}m_{1}}{20}\sigma\left(\frac{20a_{1}}{\kappa_{1}m_{1}}l\right)\right\} dl.$$

# OUTLINE

- Types of Relays
- System Model and Reference Trajectories
- Error Dynamics and Objective
- Stability Theorem
- Crafting the Reference Trajectory
- Numerical Validation
- Conclusions

Reference Trajectory



Reference Trajectory



Mimics square wave with .01 offset.

$$y_d(t) = 0.01 + \varepsilon_1 \left[ \mathcal{I}(500 + \min\{t, 50\}) - \mathcal{I}(\min\{\max\{t, 450\}, 550\}) + \mathcal{I}(\max\{t, 950\} - 500) \right] \text{ for } 0 \le t \le 1000,$$
  
$$y_d(t) = y_d(t - 1000) \text{ for } t \ge 1000,$$

where

$$\mathcal{I}(r) = \int_{450}^{r} (s - 450)^3 (550 - s)^3 \mathrm{d}s$$

and

 $\varepsilon_1 = .99/\mathcal{I}(550).$ 

Reference Trajectory



Mimics square wave with .01 offset.

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NUMERICAL VALIDATION

We chose

$$a_1 = a_2 = 1,$$
  
 $a_3 = 100,$   
 $m = 1, \quad k = 2.5, \quad \gamma = 1,$   
 $b = 1, \quad \alpha = 0.5, \quad \beta = 0.001,$   
 $g_o = 1,$   
 $(x, \dot{x}, z)(0) = (0, 0, 10).$ 





Control Signal







# OUTLINE

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- For the proofs, see [MM., F. Mazenc, and M. de Queiroz, "Tracking and robustness analysis for controlled microelectromechanical relays," *Intl. J. Robust Nonlinear Control*, to appear.]

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- For a .pdf with these slides, email malisoff at lsu dot edu.